



STK672-060

Stepping Motor Driver (Sine Wave Drive) Output Current: 1.2 A (No Heat Sink*)

Unipolar constant-current chopper (external excitation PWM) circuit with built-in microstepping controller

Overview

The STK672-060 is a stepping motor driver hybrid IC that adopts power MOSFETs in its output stage. It features a built-in microstepping controller that implements unipolar constant current PWM drive. Since this IC includes a 4-phase distribution controller for stepping motors, it can contribute not only to system simplification but to circuit standardization as well.

The STK672-060 supports the 2 phase, 1-2 phase, W1-2 phase, 2W1-2 phase, and 4W1-2 phase excitation (drive) methods and can control the motor with the basic stepping angle of the stepping motor divided into up to 16 divisions. Motor rotation can also be controlled with just the clock signal.

This hybrid IC can implement highly efficient drive with high motor torque, low vibration, low noise, and fast response. As compared to the earlier Sanyo STK672-010 series, the STK672-060 features a smaller package, fewer required external components, and improvements to the controller for high functionality high performance microstepping drive.

Applications

- Facsimile unit stepping motor drive for both transmission and reception
- Paper feed and optical system stepping motor drive in copiers
- Laser printer drum drive
- Printer carriage stepping motor drive
- X/Y plotter pen drive
- Other stepping motor applications

Features

- The STK672-060 can implement a stepping motor drive system with just the provision of a DC power supply and a clock pulse generator.

<Control Block Features>

The excitation mode settings (M1, M2, and M3) select one of five excitation methods.

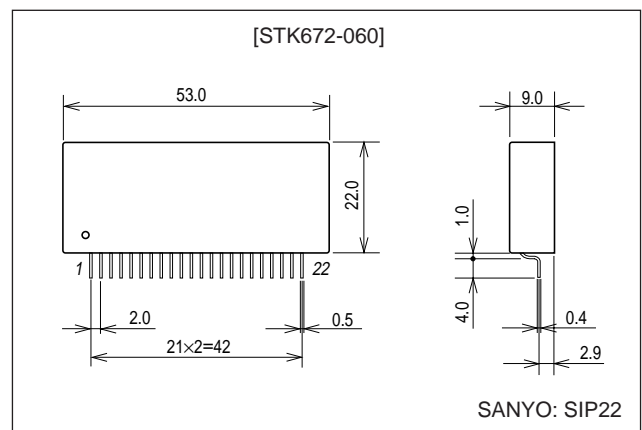
- 2 phase excitation
- 1-2 phase excitation
- W1-2 phase excitation
- 2W1-2 phase excitation
- 4W1-2 phase excitation

Continued on next page.

Package Dimensions

unit: mm

4161-SIP22



Notes*: Conditions: $V_{CC1} = 24\text{ V}$, $I_{OH} = 1.2\text{ A}$, 2W1-2 excitation mode

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- To match the motor characteristics, the vector locus during microstepping drive can be selected to be any one of four modes: circular mode, one inside mode, and two outside modes.
 - The phase is retained if excitation is switched during operation.
 - The excitation phase state can be verified in real time using the MO1, MO2, and MOI signal output pins.
 - The clock input counter block supports two signalsense modes selected by the high or low state of the M3 pin.
 - Rising edges only
 - Both rising and falling edges
 - The CLK and RETURN input pins have built-in circuits that prevent malfunctions due to external noise pulses.
 - Both an ENABLE and a RESET pin are provided as Schmitt trigger inputs with built-in 20 k Ω (typical) pull-up resistors.
 - No audible noise is generated by the differences in the time constant between phases A and B when the motor position is held fixed due to the adoption of external excitation.
 - The reference voltage Vref can be set to any level between 0 and 1/2 V_{CC2}. This allows the STK672-060 to provide microstepping operation even for small motor currents.
- <Drive Block>
- Provides a wide range of operating supply voltage required for external excitation PWM drive (V_{CC1} = 10 to 45 V).
 - Current detection resistor (0.22 Ω) built-in the hybrid IC itself.
 - Power MOSFETs adopted for low drive loss.
 - Provides a motor output drive current of I_{OH} = 1.2 A.

Specifications

Absolute Maximum Ratings at T_c = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage 1	V _{CC1} max	No signal	52	V
Maximum supply voltage 2	V _{CC2} max	No signal	−0.3 to +7.0	V
Input voltage	V _{IN} max	Logic input pins	−0.3 to +7.0	V
Phase output current	I _{OH} max	0.5 s, 1 pulse, when V _{CC1} applied	1.6	A
Repeated avalanche resistance	E _{ar} max		25	mJ
Power dissipation	P _d max	$\theta_{c-a} = 0$	7	W
Operating IC substrate temperature	T _c max		105	°C
Junction temperature	T _j max		150	°C
Storage temperature	T _{stg}		−40 to +125	°C

Allowable Operating Ranges at T_a = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage 1	V _{CC1}	When the input signal is present	10 to 45	V
Supply voltage 2	V _{CC2}	When the input signal is present	5 \pm 5%	V
Input voltage	V _{IH}		0 to V _{CC2}	V
Phase drive voltage handling capacity	V _{DSS}	Transistors 1, 2, 3, and 4 (outputs A, \bar{A} , B, and \bar{B})	100 (min)	V
Phase current	I _{OH} max	Duty 50%	1.2 (max)	A

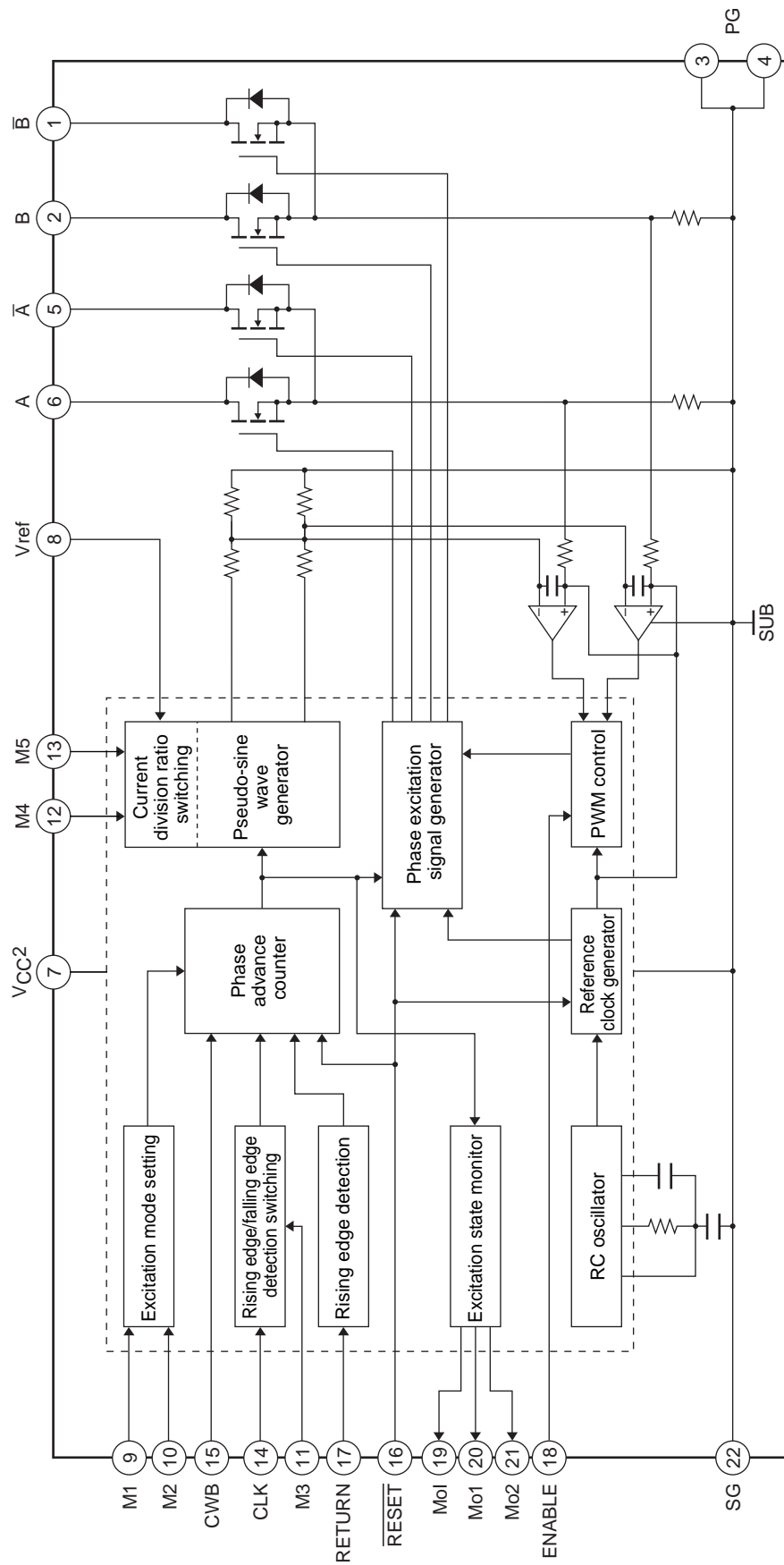
STK672-060

Electrical Characteristics at $T_c = 25^\circ\text{C}$, $V_{CC1} = 24\text{ V}$, $V_{CC2} = 5\text{ V}$

Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Control power supply current	I_{CC}	The hybrid IC pin 7 input, ENABLE = low		2.5	14	mA
Output saturation voltage	V_{sat}	$R_L = 23\ \Omega$ ($I \approx 1\text{ A}$)		0.8	1.1	V
Average output current	$I_{o\ ave}$	Load: $R = 3.5\ \Omega$. $L = 3.8\text{ mH}$ per phase, $V_{ref} \approx 1.69\text{ V}$	0.470	0.524	0.580	A
FET diode forward voltage	V_{df}	$I_f = 1\text{ A}$		1.2	1.8	V
[Control Input Pins]						
Input voltage	V_{IH}	Except for the V_{ref} pin	4			V
	V_{IL}	Except for the V_{ref} pin			1	V
Input current	I_{IH}	Except for the V_{ref} pin	0	1	10	μA
	I_{IL}	Except for the V_{ref} pin	125	250	510	μA
[Vref Input Pin]						
Input voltage	V_I	H-IC pin 8	0		2.5	V
Input current	I_I	H-IC pin 8, $V_I = 2.5\text{ V}$	330	415	545	μA
[Control Output Pins]						
Output voltage	V_{OH}	$I = -3\text{ mA}$, Mol, Mo1, Mo2 pins	2.4			V
	V_{OL}	$I = +3\text{ mA}$, Mol, Mo1, Mo2 pins			0.4	V
[Current Division Ratio (A • B)]						
2W1-2, W1-2, 1-2	V_{ref}	$\theta = 1/8$		100		%
2W1-2, W1-2	V_{ref}	$\theta = 2/8$		92		%
2W1-2	V_{ref}	$\theta = 3/8$		83		%
2W1-2, W1-2, 1-2	V_{ref}	$\theta = 4/8$		71		%
2W1-2	V_{ref}	$\theta = 5/8$		55		%
2W1-2, W1-2	V_{ref}	$\theta = 6/8$		40		%
2W1-2	V_{ref}	$\theta = 7/8$		21		%
2	V_{ref}			100		%
PWM frequency	f_c		37	47	57	kHz

Notes: A constant voltage power supply must be used.
Design target values are shown for the current division ratios.

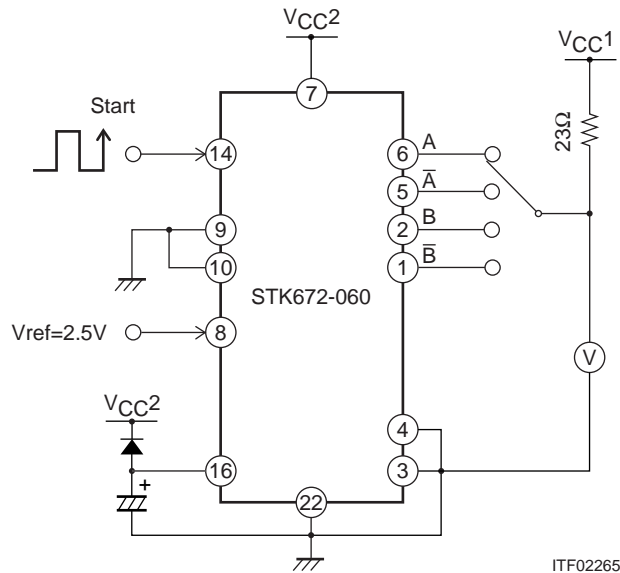
Block Diagram



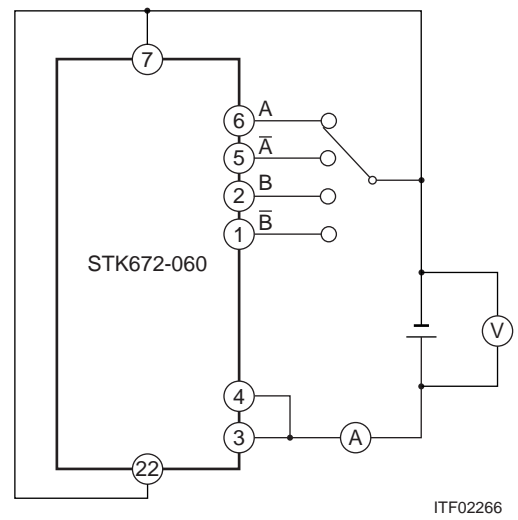
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Test Circuit Diagrams

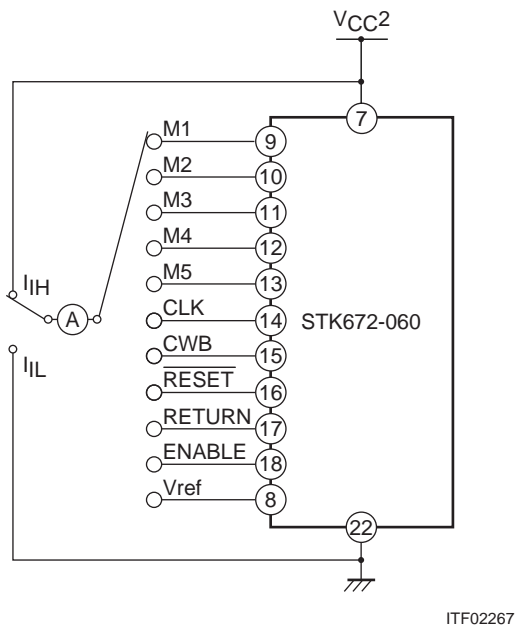
V_{sat}



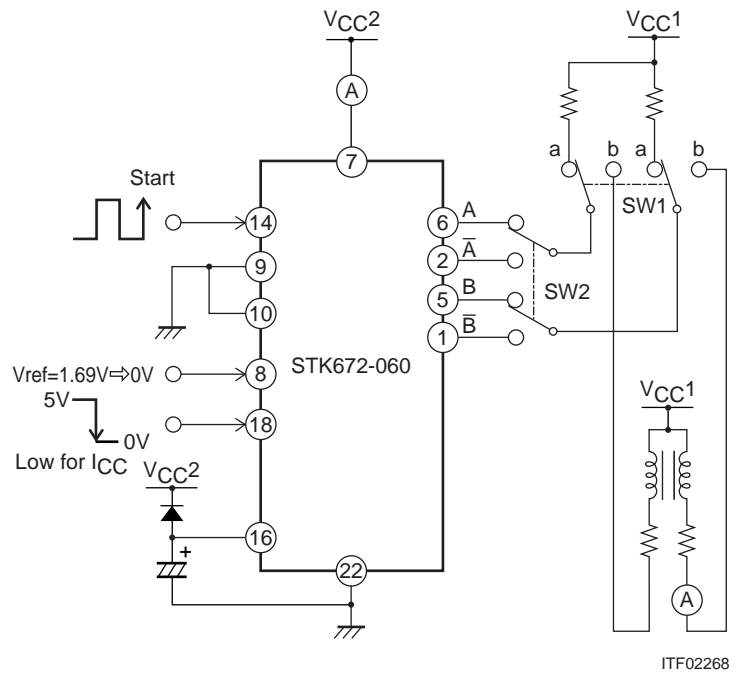
V_{df}



I_{IH}, I_{IL}



I_{oave}, I_{cc}, f_c



When measuring I_{oave}: Set switch SW1 to the b position, set V_{ref} = 1.69 V
 When measuring f_c: Set switch SW1 to the a position, set V_{ref} = 0 V
 When measuring I_{cc}: Set ENABLE low.

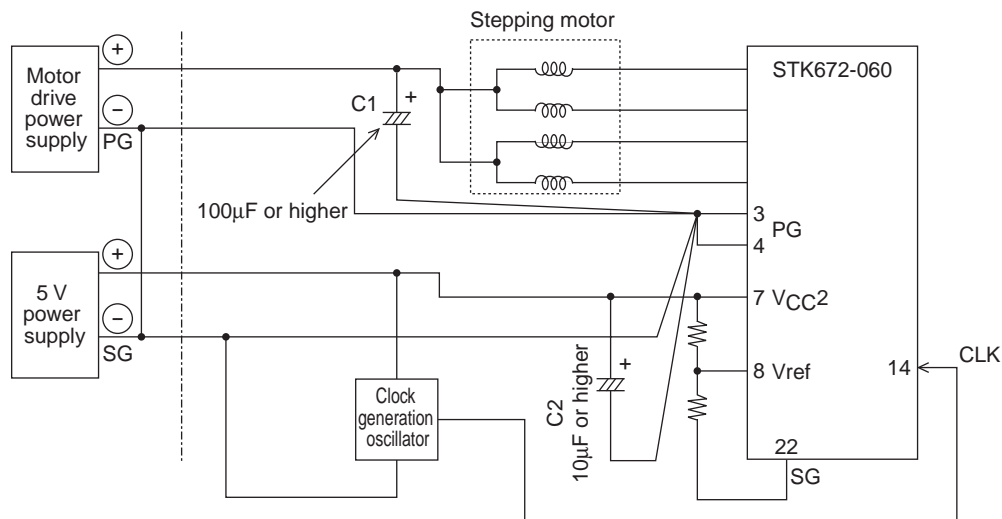
Recommendations for PCB Designs

This hybrid IC has three ground pins: the PG pins (pins 3 and 4) and the SG pin (pin 22). These pins are connected internally.

Two power supplies are required: one power supply for motor drive and another for the hybrid IC 5 V control system. If there are problems with the ground connections for these power supplies, that can in turn cause the motor current waveform to become unstable, an increase in audible motor noise, or increased motor vibration. The ground lines must be designed appropriately.

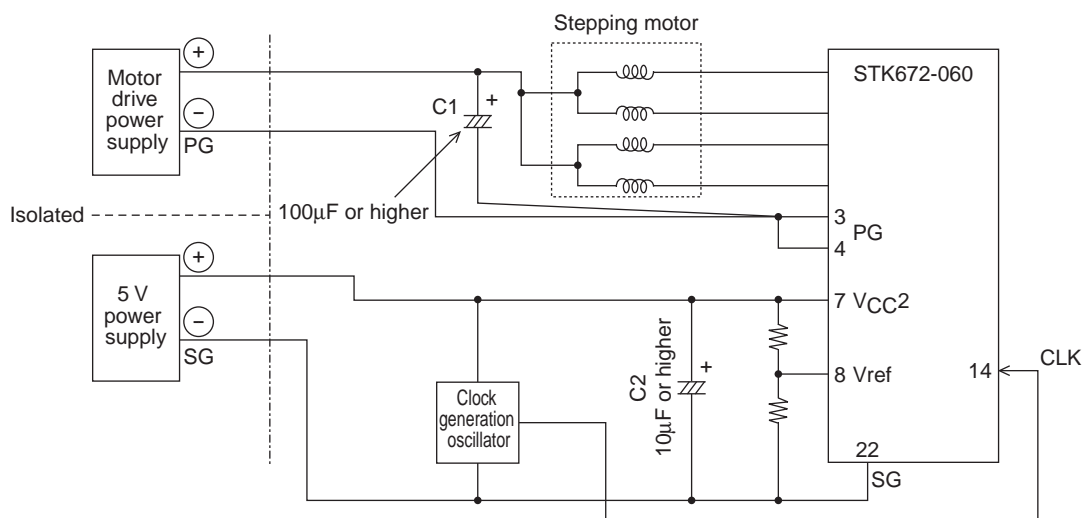
This section describes two ground connection methods.

- When the grounds for the motor drive power supply and the hybrid IC 5 V power supply are connected close to the power supplies
 - a) If PG and SG are shorted at the power supply side, connect only the PG line to pins 3 and 4 on the hybrid IC. Verify that voltage drops due to common line impedances do not occur. Note that V_{CC2} is required to be within $\pm 5\%$ in the specifications.
 - b) Connecting the Vref ground to pin 22 provides more stable current waveforms.
 - c) For initial values, use 100 μF or higher for C1 and 10 μF or higher for C2.
C1 must be located close to the hybrid IC and the capacitor ground line must be as short as possible.



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- When the grounds for the motor drive power supply and the hybrid IC 5 V power supply are separated
 - a) Place the capacitor C1 (100 μF or higher) as close to the hybrid IC as possible. This capacitor's ground line must be as short as possible.
Connect the capacitor C2 of an appropriate value if required. Its ground line must be as short as possible.



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• Input Signal Functions and Timing

CLK (Phase switching clock)

Input frequency range: DC to 50 kHz

Minimum pulse width: 10 μ s

Duty: 40 to 60% (The minimum pulse width takes precedence when M3 is high)

Pin circuit type: CMOS Schmitt trigger with built-in pull-up resistor (20 k Ω typical)

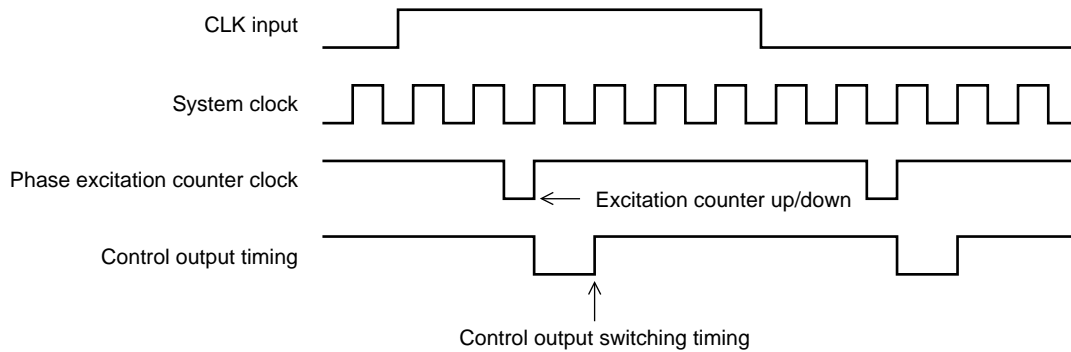
A multi-stage noise filter is built in.

Function

When M3 is high or open: The excited phase is advanced by one step on each rising edge of CLK signal.

When M3 is low: The phase is advanced 2 steps on the rising and falling edges of the CLK signal.

CLK Input Acquisition Timing (when M3 is low)



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CWB (Rotation setting procedure)

Pin circuit type: CMOS Schmitt trigger with built-in pull-up resistor (20 k Ω typical)

Function

When CWB is low: Rotation in the clockwise direction.

When CWB is high: Rotation in the counterclockwise direction

Note: When M3 is low, the CWB input must not be changed within $\pm 6.25 \mu$ s of a rising or falling edge on the CLK input.

RETURN (Forcible return to the origin point for the current phase)

Pin circuit type: CMOS Schmitt trigger with built-in pull-up resistor (20 k Ω typical)

Built-in noise filter

Note: The motor is forcibly moved to the origin point for the current phase by changing the input level on this pin from low to high. When unused, this pin must normally be left open or connected to the V_{CC2} .

ENABLE (On/off control of the A, \bar{A} , B, and \bar{B} excitation drive outputs and selection of the internal operate or hold state of the hybrid IC itself)

Pin circuit type: CMOS Schmitt trigger with built-in pull-up resistor (20 k Ω typical)

Function

When ENABLE is high or open: Normal operating state

When ENABLE is low: The hybrid IC goes to the hold state, and the excitation drive output (motor current) is forcibly turned off (the output current is cut off). In this state, the hybrid IC system clock is stopped, and the hybrid IC is not affected by any changes in the state of the input pins other than the reset input.

M1, M2, and M3 (Excitation mode and CLK input edge timing selection)

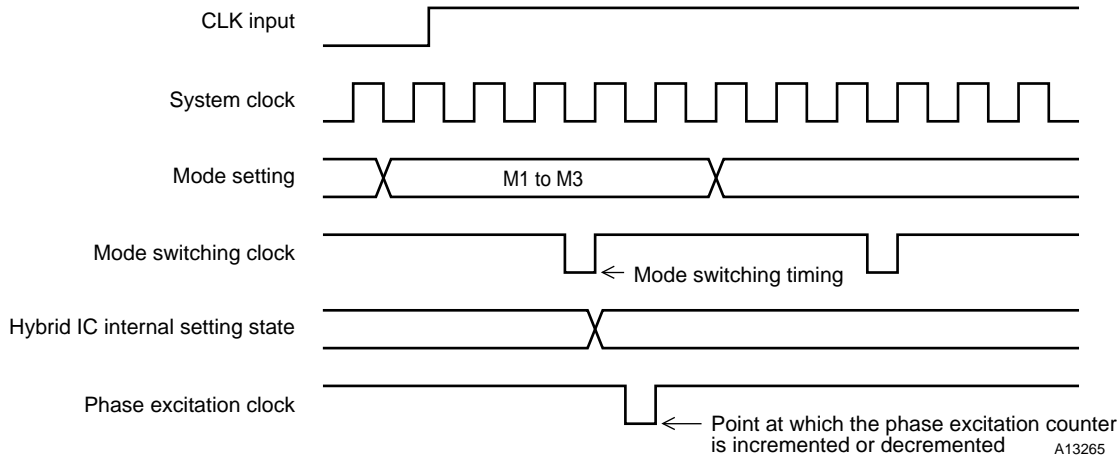
Pin circuit type: CMOS Schmitt trigger with built-in pull-up resistor (20 kΩ typical)

Function

M2	0	0	1	1	Phase switching CLK edge timing
M1	0	1	0	1	
M3	0	1	0	1	
1	2 phase excitation	1-2 phase excitation	W1-2 phase excitation	2W1-2 phase excitation	Rising edge only
0	1-2 phase excitation	W1-2 phase excitation	2W1-2 phase excitation	4W1-2 phase excitation	Both rising and falling edges

Timing when mode setting is enabled: Do not change the mode within $\pm 5 \mu\text{s}$ of a CLK signal rising or falling edge.

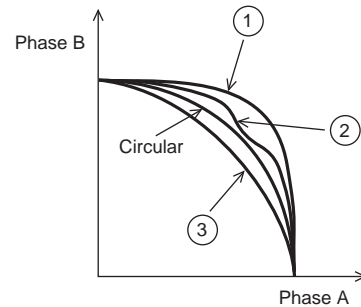
Mode Setting Acquisition Timing



M4 and M5 (Rotation vector locus setting for microstepping mode)

M4	1	0	1	0
M5	1	0	0	1
Mode	Circular	①	②	③

See the table on the following page for details on the current division ratio.



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RESET (Resets the whole system)

Pin circuit type: CMOS Schmitt trigger with built-in pull-up resistor (20 kΩ typical)

Function: The circuit states are all set to their initial values by applying a low level (with a pulse width of 10 μs or longer) to the **RESET** pin. At this time, the A and B phases are set to their origin points, regardless of the excitation mode. The output current becomes about 71% after the reset is cleared.

Note: The Vref voltage is established by applying a reset after power is first applied. Applications must perform a power-on reset operation after the VCC2 power is applied.

Vref (Sets the current value used as the reference for constant-current output)

Pin circuit type: Analog input circuit

Function: Applications can implement constant-current control of the motor excitation current at 100% of the rated current value by applying a voltage lower than the control system power supply voltage VCC2 minus 2.5 V. This IC supports constant current control proportional to the Vref voltage with an upper limit of 2.5 V.

Output Pin Description

Pin No.	Pin	Function	Pin circuit type
19	MoI	Phase excitation origin point monitor	Standard CMOS output
20, 21	Mo1, Mo2	Phase excitation status monitor	Standard CMOS output

Output Signal Functions and Timings

A, \bar{A} , B, and \bar{B} (Motor phase excitation outputs)

Function: In 4 phase and 2 phase excitation modes, an interval of about 3.75 μ s (typical) is set up when the A and \bar{A} , and B and \bar{B} output signals change their state.

MO1, MO2, and MOI (Motor phase excitation state monitoring)

Pin circuit type: Standard CMOS output

Function: These pins output the current phase excitation output state.

Phase coordinate	Phase A	Phase B	Phase \bar{A}	Phase \bar{B}
Mo1	1	0	0	1
Mo2	0	1	0	1

MoI outputs a 0 when the corresponding phase is at its origin, and a 1 at all other positions.

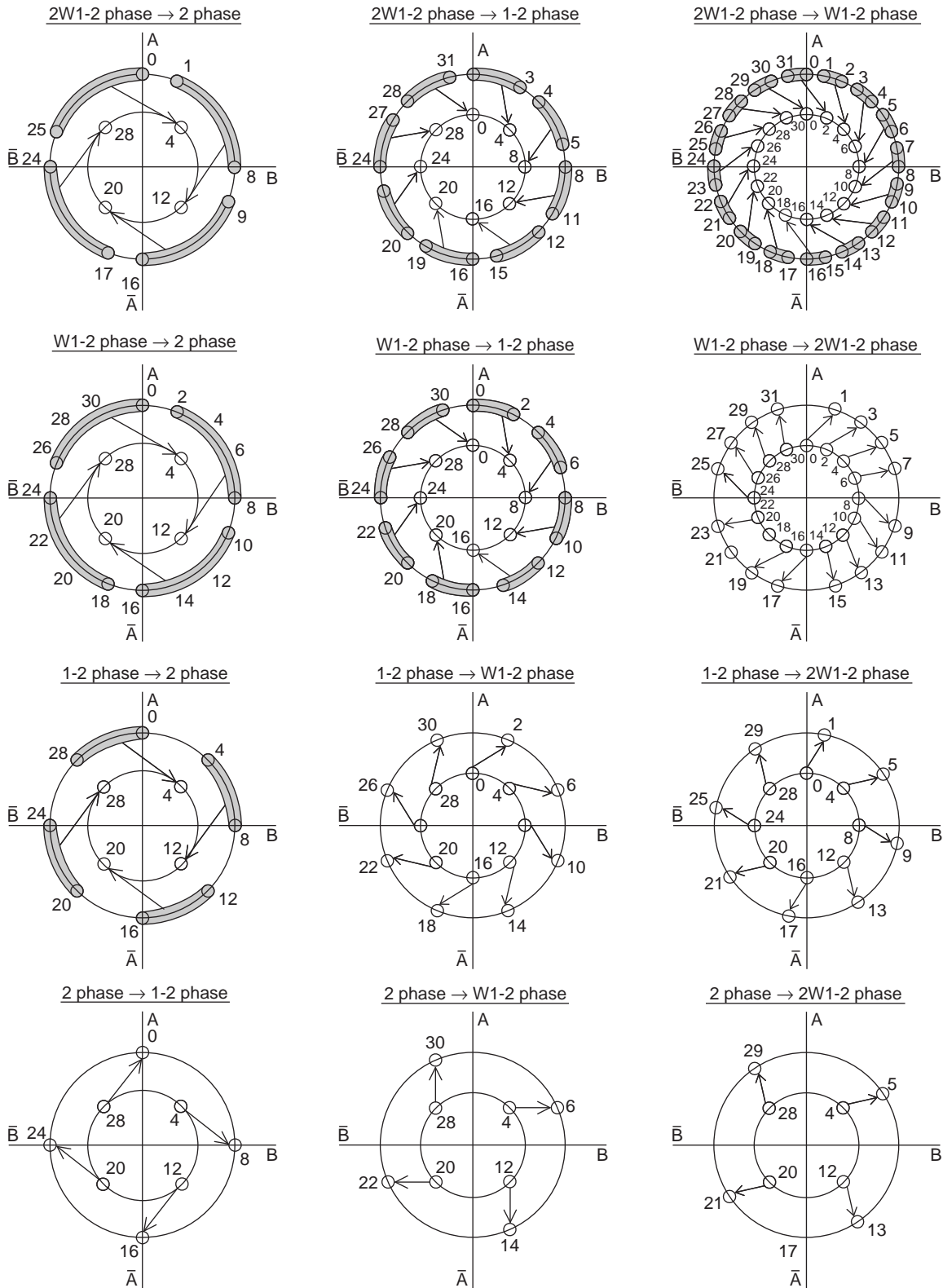
Current Division Ratios as Set by M3, M4, and M5 Values provided for reference purposes

Mode			Circular	①	②	③	Unit	Number of steps	
Setting	M3 = 0	M3 = 1	M4 = 1 M5 = 1	M4 = 0 M5 = 0	M4 = 1 M5 = 0	M4 = 0 M5 = 1			
Current division ratio	4W1-2		15	16	16	15	%		1/16
		2W1-2	21	25	24	20		1/8	2/16
			31	34	33	28			3/16
		2W1-2	40	44	41	38		2/8	4/16
			47	50	49	44			5/16
		2W1-2	55	59	56	53		3/8	6/16
			63	67	63	60			7/16
		2W1-2	71	75	70	67		4/8	8/16
			76	81	76	73			9/16
		2W1-2	83	87	84	81		5/8	10/16
			87	92	88	84			11/16
		2W1-2	92	95	95	91		6/8	12/16
			96	98	98	93			13/16
		2W1-2	100	100	100	100		7/8	14/16

Load conditions: $V_{CC1} = 24$ V, $V_{CC2} = 5$ V, $R/L = 3.5$ Ω /3.8 mH

Phase States at Excitation Switching

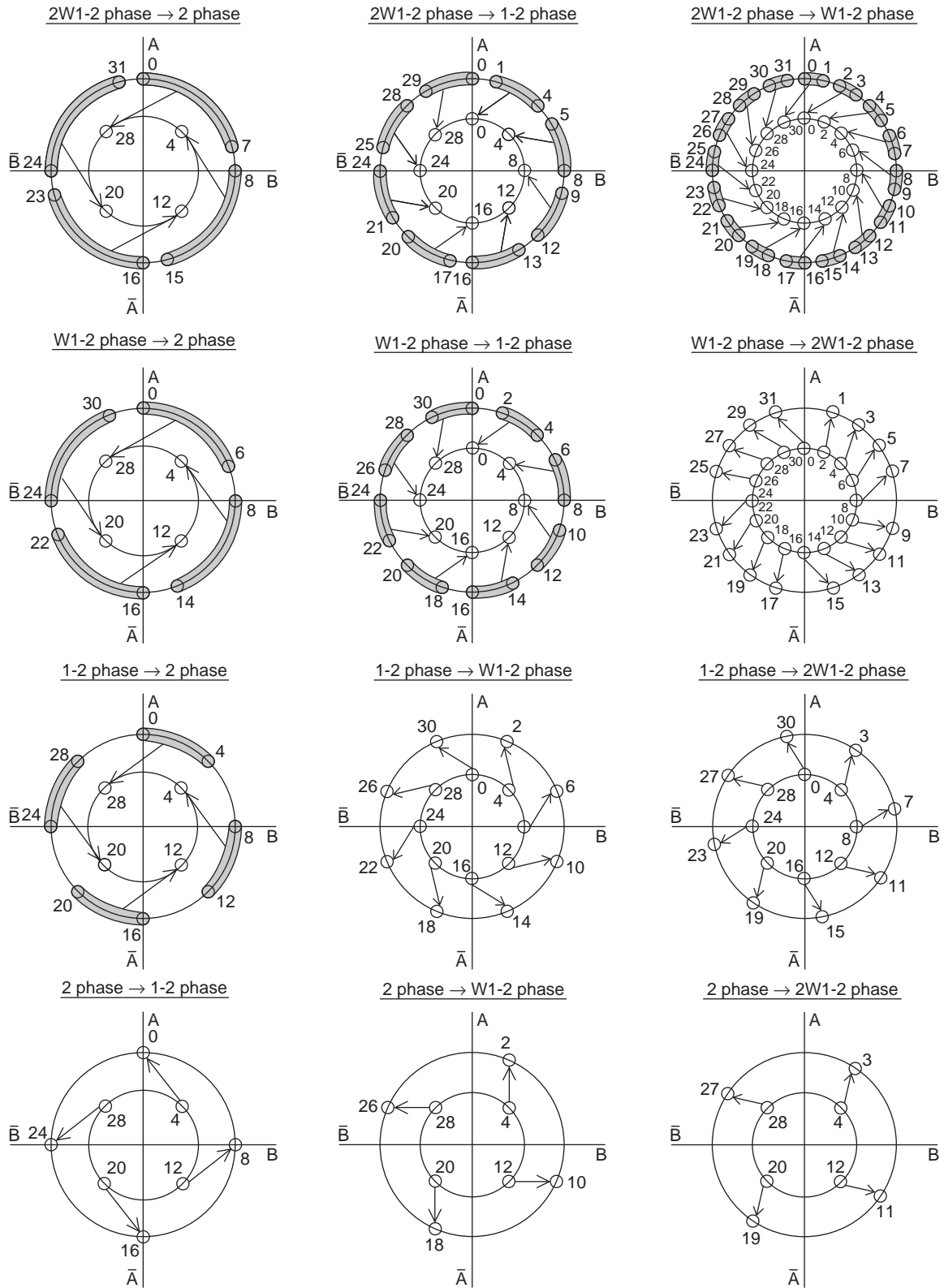
Excitation Phases Before and After Switching Excitation Modes <Clockwise direction>



Excitation phase developed by the first CLK internal pulse after the change in the excitation mode setting with M1 and M2
Excitation phase immediately prior to the excitation mode setting

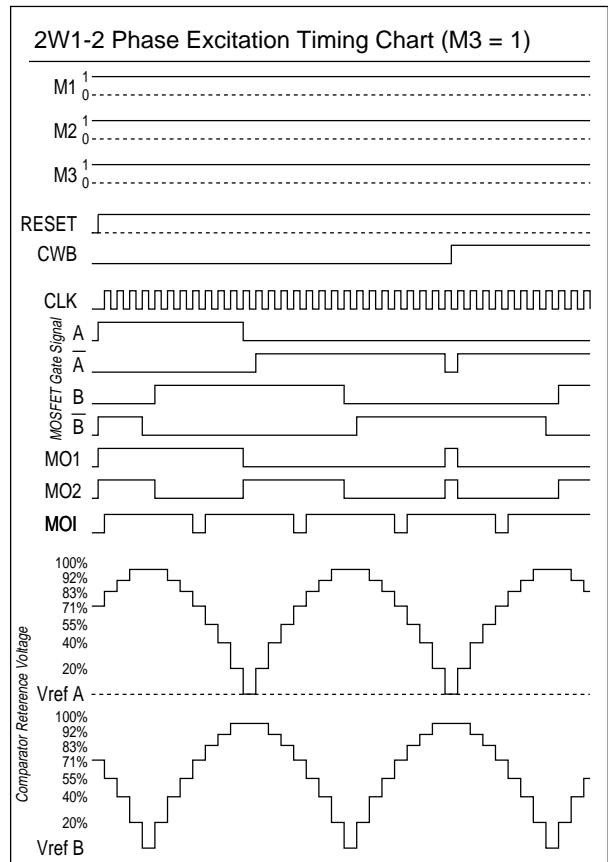
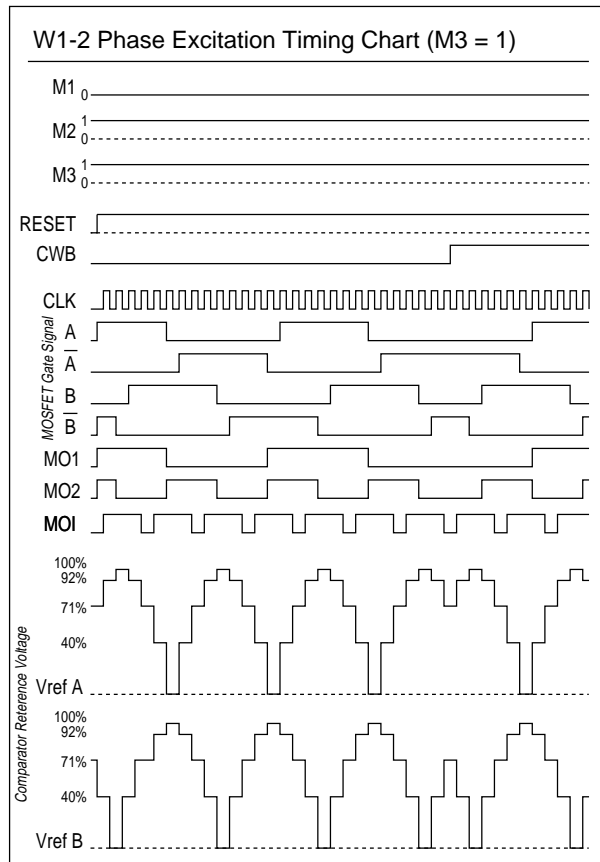
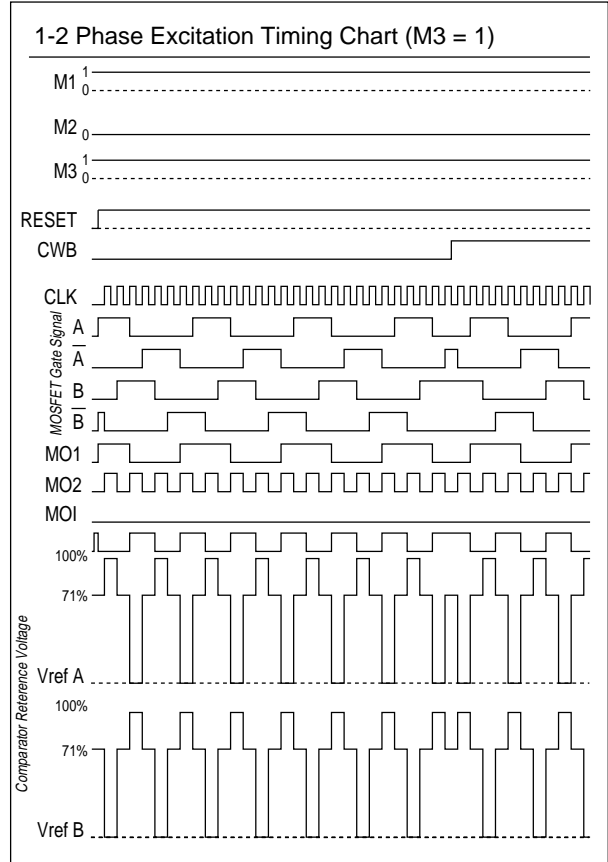
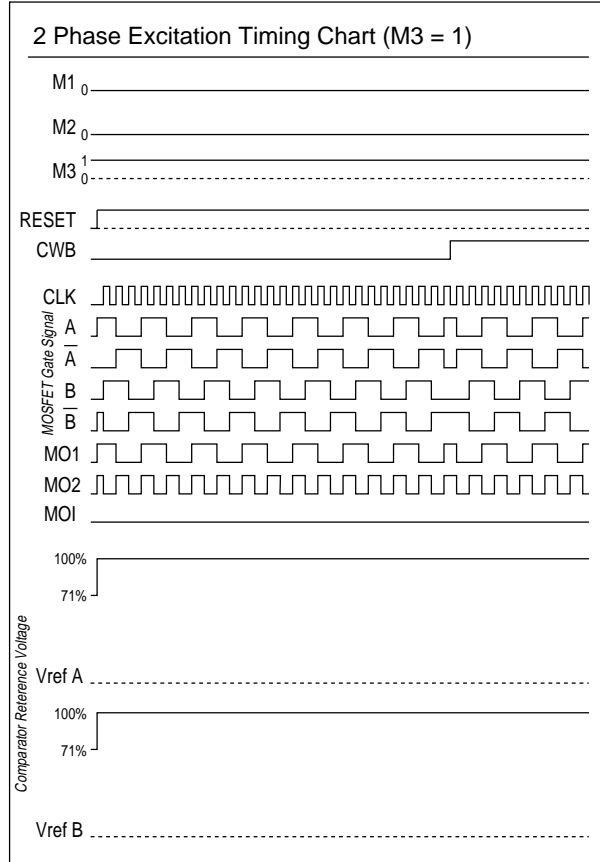
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Excitation Phases Before and After Switching Excitation Modes <Clockwise direction>



Excitation Time and Timing Charts

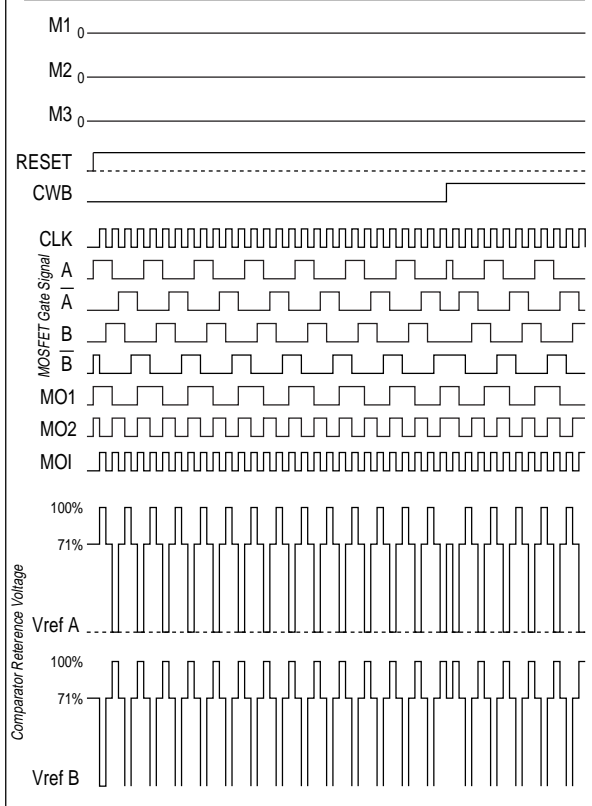
CLK Rising Edge Operation



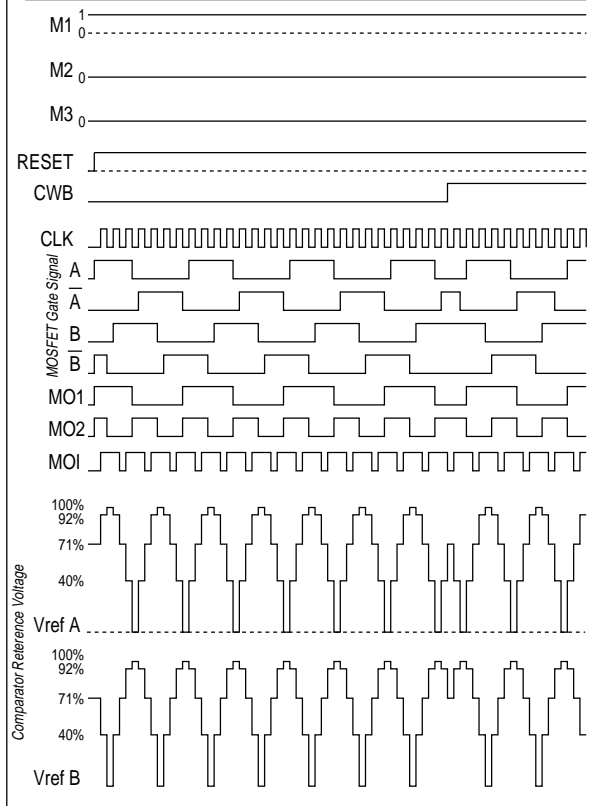
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CLK Rising and Falling Edge Operation

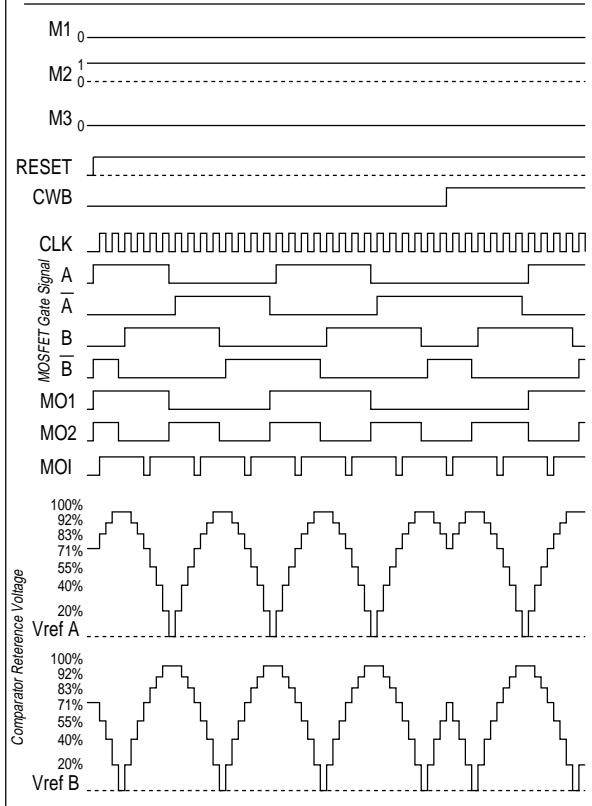
1-2 Phase Excitation Timing Chart (M3 = 0)



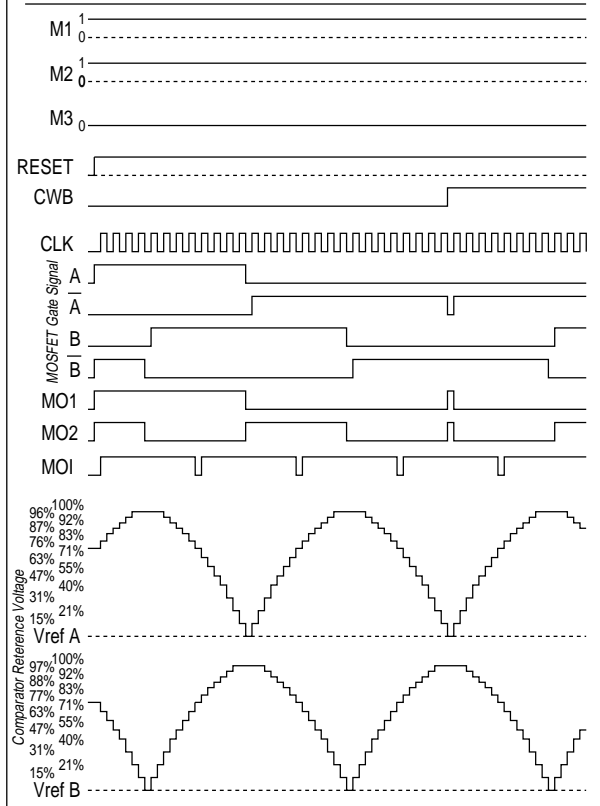
W1-2 Phase Excitation Timing Chart (M3 = 0)



2W1-2 Phase Excitation Timing Chart (M3 = 0)



4W1-2 Phase Excitation Timing Chart (M3 = 0)



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Thermal Design

<Hybrid IC Internal Average Power Dissipation, Pd>

The devices with the largest average power dissipation levels in this hybrid IC are the current control devices, the regenerative current diodes, and the current detection resistor. Since sine wave drive is used, the average power dissipation associated with microstepping drive can be approximated by applying a waveform ratio (0.64 in this case) to square wave power dissipation in 2-phase excitation.

The power dissipation levels for the available excitation modes are listed below.

$$\text{2 phase excitation} \quad Pd_{2EX} = (V_{sat} + V_{df}) \cdot \frac{f_{clock}}{2} \cdot I_{OH} \cdot t_2 + \frac{I_{OH} \cdot f_{clock}}{2} \cdot (V_{sat} \cdot t_1 + V_{df} \cdot t_3)$$

$$\text{1-2 phase excitation} \quad Pd_{1-2EX} = 0.64 \cdot \left\{ (V_{sat} + V_{df}) \cdot \frac{f_{clock}}{4} \cdot I_{OH} \cdot t_2 + \frac{I_{OH} \cdot f_{clock}}{4} \cdot (V_{sat} \cdot t_1 + V_{df} \cdot t_3) \right\}$$

$$\text{W1-2 phase excitation} \quad Pd_{W1-2EX} = 0.64 \cdot \left\{ (V_{sat} + V_{df}) \cdot \frac{f_{clock}}{8} \cdot I_{OH} \cdot t_2 + \frac{I_{OH} \cdot f_{clock}}{8} \cdot (V_{sat} \cdot t_1 + V_{df} \cdot t_3) \right\}$$

$$\text{2W1-2 phase excitation} \quad Pd_{2W1-2EX} = 0.64 \cdot \left\{ (V_{sat} + V_{df}) \cdot \frac{f_{clock}}{16} \cdot I_{OH} \cdot t_2 + \frac{I_{OH} \cdot f_{clock}}{16} \cdot (V_{sat} \cdot t_1 + V_{df} \cdot t_3) \right\}$$

$$\text{4W1-2 phase excitation} \quad Pd_{4W1-2EX} = 0.64 \cdot \left\{ (V_{sat} + V_{df}) \cdot \frac{f_{clock}}{16} \cdot I_{OH} \cdot t_2 + \frac{I_{OH} \cdot f_{clock}}{16} \cdot (V_{sat} \cdot t_1 + V_{df} \cdot t_3) \right\}$$

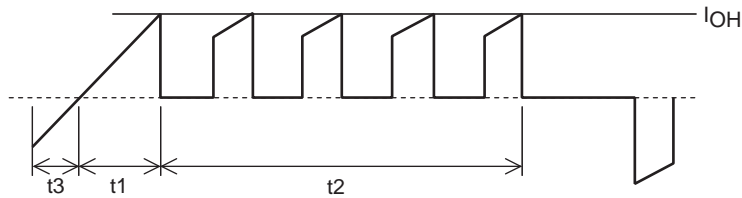
The values of t1 and t3 can be derived from the same formula for each excitation mode

$$t_1 = \frac{-L}{R + 0.7} \cdot \ln \left(1 - \frac{R + 0.7}{V_{CC1}} \cdot I_{OH} \right) \quad t_3 = \frac{-L}{R} \cdot \ln \left(\frac{V_{CC1} + 0.7}{I_{OH} \cdot R + V_{CC1} + 0.7} \right)$$

The formula for t2 differs for the excitation modes.

$$\text{2 phase excitation} \quad t_2 = \frac{2}{f_{clock}} - (t_1 + t_3) \quad \text{1-2 phase excitation} \quad t_2 = \frac{3}{f_{clock}} - t_1$$

$$\text{W1-2 phase excitation} \quad t_2 = \frac{7}{f_{clock}} - t_1 \quad \begin{matrix} \text{2W1-2 phase excitation} \\ \text{4W1-2 phase excitation} \end{matrix} \quad t_2 = \frac{15}{f_{clock}} - t_1$$



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Motor Phase Current Model (2 phase excitation)

fclock: The CLK pin input frequency (Hz)
Vsat: The voltage drop (V) across the power MOSFET and the current detection resistor
Vdf: The voltage drop (V) across the diode and the current detection resistor
IOH: Phase current peak-to-peak value
t1: Phase current rise time (s)
t2: Constant-current operating time (s)
t3: Phase switchover current regeneration time (s)
VCC1: Supply voltage applied to the motor (V)
L: Motor inductance (H)
R: Motor winding resistance (Ω)

<Determining the Heat Sink Size for this Hybrid IC>

First, we determine the heat sink thermal resistance θ_{c-a} from the average power dissipation determined in the previous section.

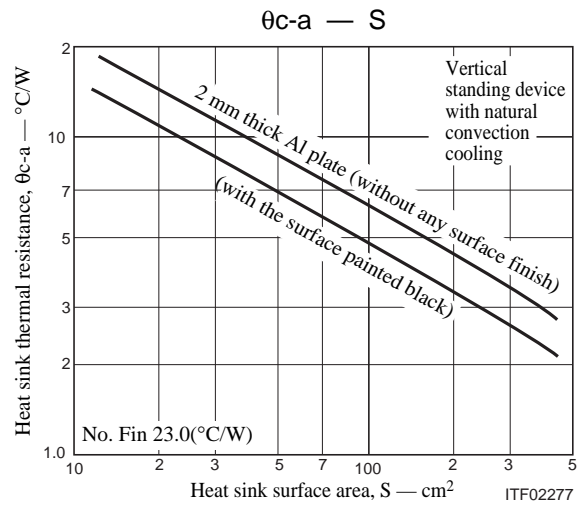
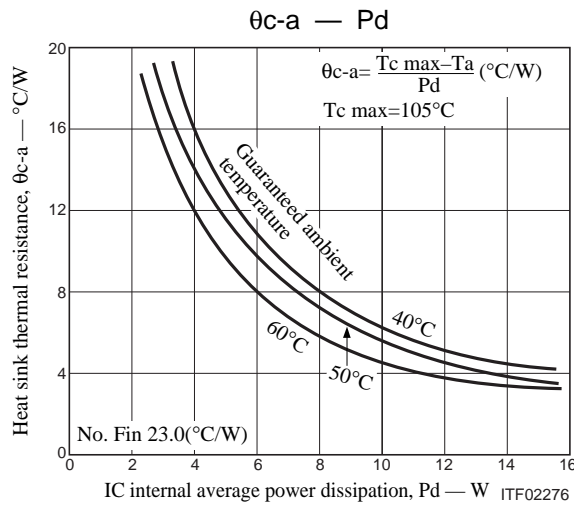
$$\theta_{c-a} = \frac{T_{c \max} - T_a}{P_{dEX}} \quad [^{\circ}\text{C/W}]$$

$T_{c \max}$: Hybrid IC substrate temperature ($^{\circ}\text{C}$)

T_a : End product internal temperature ($^{\circ}\text{C}$)

P_{dEX} : Hybrid IC internal average power dissipation (W)

After determining θ_{c-a} from the above formula, determine the size S (cm^2) of the heat sink from the following graphs. The ambient temperature depends strongly on the ventilation conditions within the end product. Thus the size of the heat sink must be checked carefully with the IC installed in the end product so that the back surface (the aluminum surface) of this hybrid IC never exceeds $T_{c \max}$ (105°C) under all possible operating conditions.



Next, with the hybrid IC used without fins, we determine the allowable hybrid IC average internal power dissipation from H-IC substrate thermal resistance $\theta_{c-a} = 23^{\circ}\text{C/W}$.

Assuming that $T_{c \max}$ is 105°C at an ambient temperature of 50°C : $P_{dEX} = \frac{105 - 50}{23} = 2.3 \text{ W}$

With an ambient temperature of 40°C and a $T_{c \max}$ of 105°C : $P_{dEX} = \frac{105 - 40}{23} = 2.8 \text{ W}$

This device can be used under all operating conditions without fins provided that all dissipation values are not exceeded. (See the $\Delta T_c - P_d$ curve.)

<Hybrid IC Internal Power Device (MOSFET) Junction Temperature Calculation>

Here we determine the junction temperature T_j for each device from the power dissipation for each transistor, P_{ds} and θ_{j-c} .

$$T_j = T_c + \theta_{j-c} \times P_{ds} \quad (^{\circ}\text{C})$$

Here, to determine P_{ds} , we calculate P_{dEX} separately for each excitation mode and determine the power dissipation P_{ds} for each device.

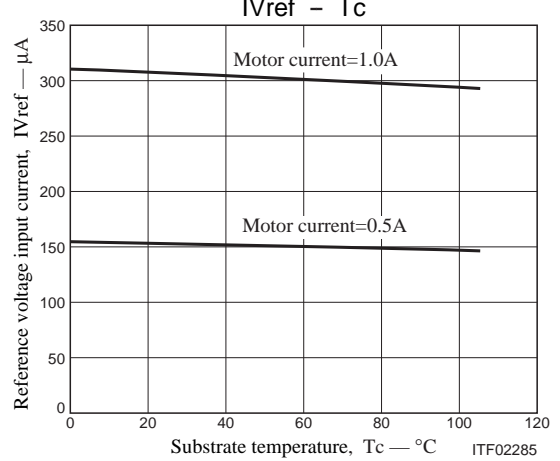
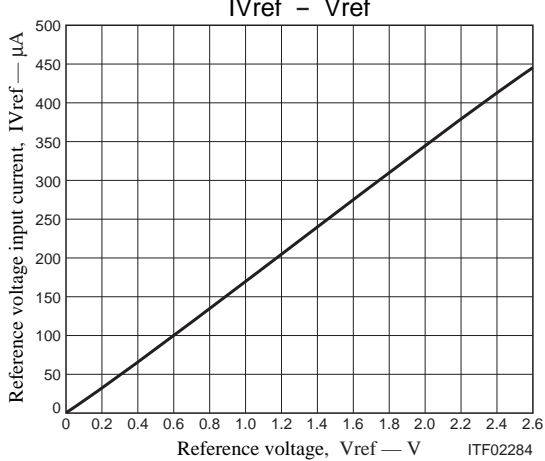
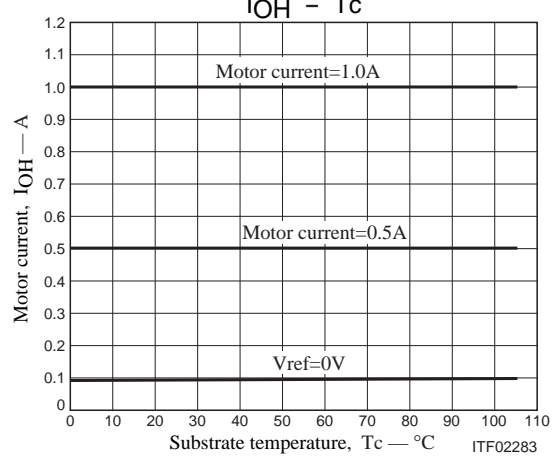
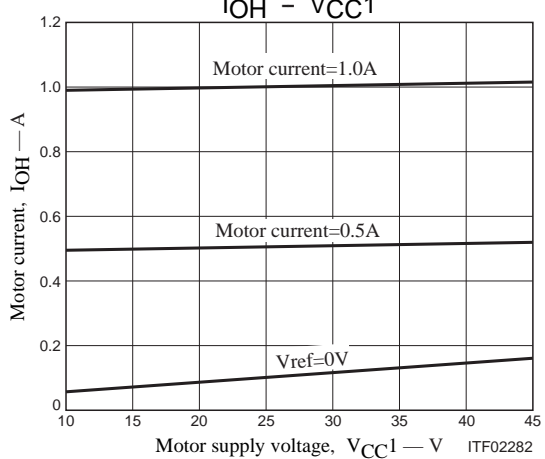
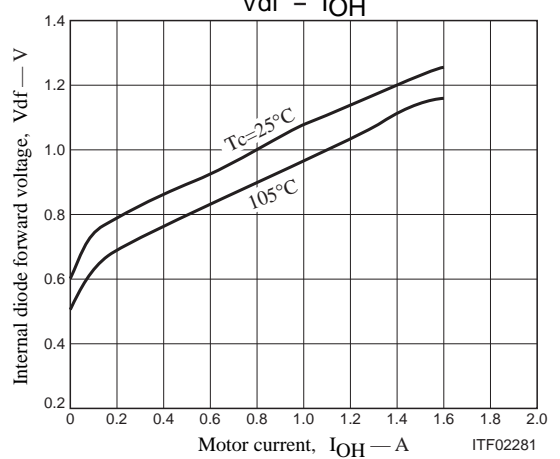
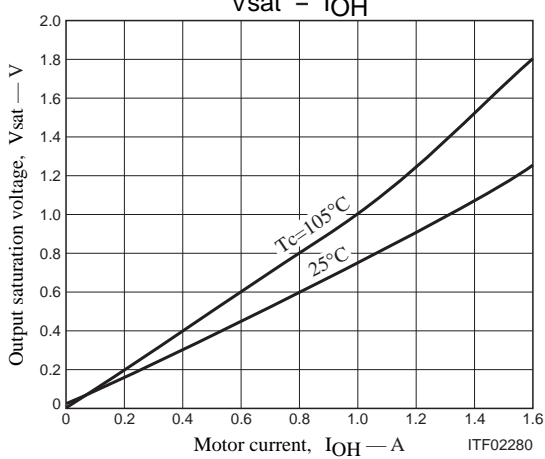
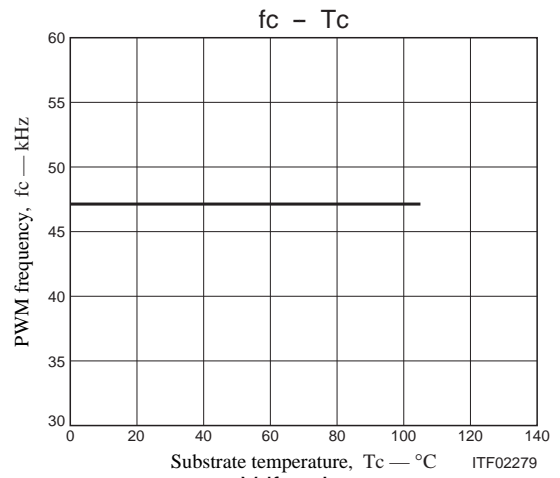
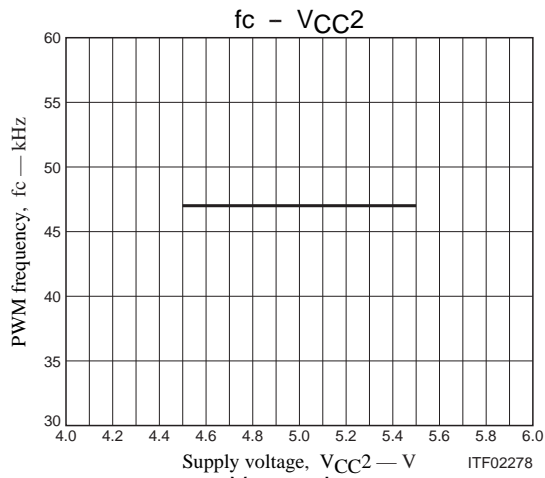
$$P_{ds} = P_d/4$$

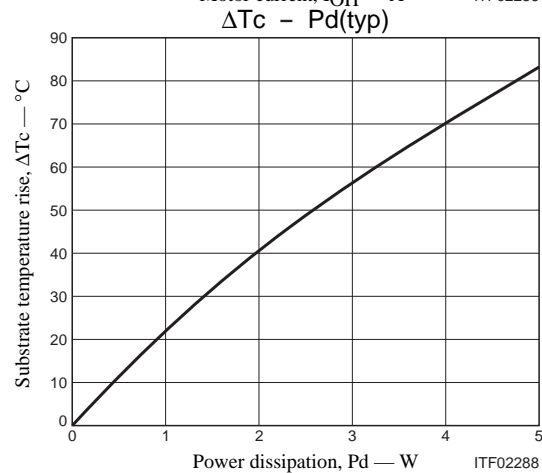
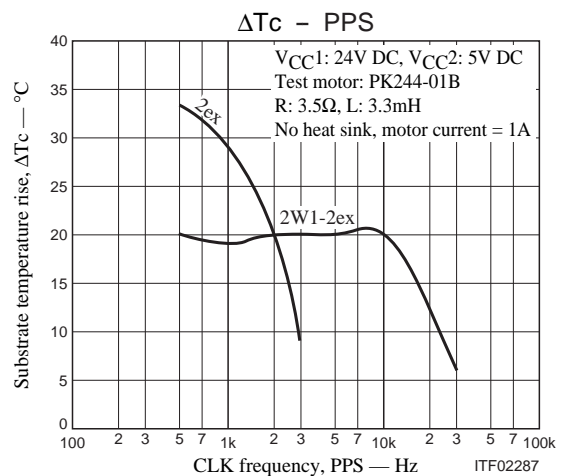
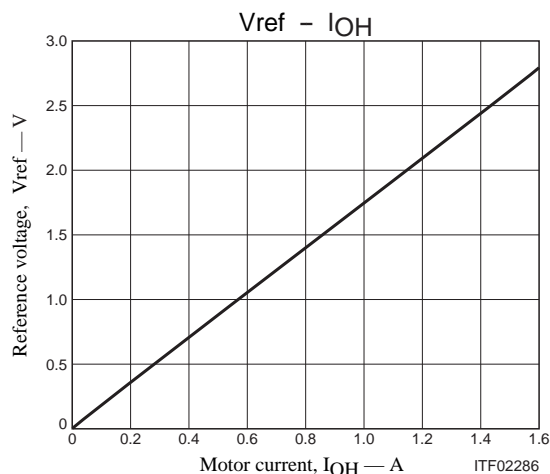
Since the power dissipation in the current detection resistor is included in the average power dissipation, here we consider that voltage drop to determine the power dissipation.

$$V_{sat} = I_{OH} \cdot R_{on} + I_{OH} \cdot R_s$$

$$V_{df} = V_{df} + I_{OH} \cdot R_s$$

The steady-state thermal resistance of the power MOSFET is 18°C/W .





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